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# DC - 1 MHz Wattmeter Based on RMS Voltage Measurements

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<u>Abstract</u> - A wideband wattmeter for measuring active power over a frequency range of dc to 1 MHz is described. The wattmeter is based on the three voltmeter method in which three rms voltage measurements are used to calculate power.

# I. INTRODUCTION.

The three voltmeter method for measuring power was described by Ayrton nearly 100 years ago. More recently, various implementations of the technique have been described [1,2]. The applied power is derived from a set of three rms voltage measurements using the law of cosines.

### **II. WATTMETER DESIGN**

A wideband wattmeter has been developed based on this principle that consists of a commercial digital voltmeter (DVM), a four-terminal resistor, two inductive voltage dividers, an isolation transformer, and associated control circuitry. This wattmeter was designed to be used with a source of synthetic power to test wattmeters over a wide range of power, power factor, and frequency. A simplified circuit diagram of the test system is shown in Fig. 1.

A test voltage V is applied to the voltage terminals of the wattmeter under test (MUT) and to inductive voltage divider  $T_1$ , which is used in conjunction with the buffer amplifier  $A_1$  to scale V to  $V_1$  (approximately 5 Vrms). The test current I is applied to the four-terminal resistor R in series with the current terminals of the test wattmeter. The voltage developed across R is converted to a ground-referenced voltage  $V_2$  (also approximately

5 Vrms) using the two-stage transformer  $T_2$  and buffer amplifier  $A_2$ . Amplifiers  $A_1$  and  $A_2$  are special composite, non-inverting amplifiers needed to minimize the loading errors of  $T_1$  and  $T_2$ . They introduce errors of less than 5 parts in 10<sup>6</sup> in the 50 Hz to 1 kHz range [3]. Difference voltage  $V_D = V_1 - V_2$  is converted to a groundreferenced voltage  $V_3$  using a center-tapped inductive voltage divider  $T_3$ , thereby avoiding the need for any active circuitry to perform the difference function. A high precision, wideband DVM is used to measure the three ground-referenced voltages  $V_1$ ,  $V_2$ , and  $V_3$ . These voltages are related to the phase angle  $\theta$  between  $V_1$  and  $V_2$  by the *Law of Cosines*:

$$V_{D}^{2} = V_{1}^{2} + V_{2}^{2} + V_{1} V_{2} \cos \theta$$

which may be expressed as:

$$V_1 V_2 \cos \theta = V_D^2 - V_1^2 - V_2^2$$
.

The magnitude of the difference voltage  $V_D$  can be described in terms of the three measured voltages and the ratio of  $T_3$  (0.5) by the following equation:

$$V_{\rm D} = [2(V_1^2 - V_2^2 - 2V_3^2)]^{\frac{1}{2}}$$

The power applied to the test wattmeter is given by:

$$P = VI\cos\theta = (r_1 r_2 V_1 V_2 \cos\theta)/R$$

where  $r_1$  is the step-down ratio of  $T_1$  and  $r_2$  is the step-up ratio of  $T_2$ .

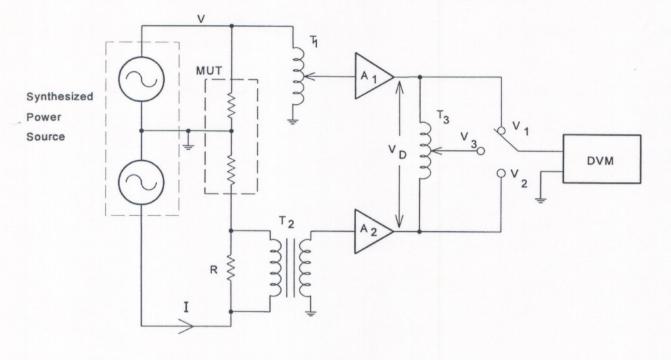


Fig. 1. Simplified diagram of the wattmeter based on rms voltage measurements.

# III. CONCLUSION

The final paper will describe the implementation of this [1] wattmeter in a system to calibrate test wattmeters that operate at 120 V and 5 A, at any power factor, over a wide frequency range with a low audio-frequency active power uncertainty of 30 parts in  $10^6$  (1- $\sigma$ ).

Standards to support precision power measurements presently are limited to the audio frequency range [4]. The wattmeter described here was developed to support the calibration of commercial power analyzers capable of [3] measuring power components out to 1 MHz. Complete performance results as well as results of comparisons with other power measurement standards will be given in the final paper. [4]

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